A NOVEL CLASS OF FAST ADAPTIVE ALGORITHMS FOR MULTICHANNEL FILTERING by

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Abstract. The major computational burden in fast transversal adaptive schemes is contributed from the time updates of the associated forward and backward predictors. In this paper the order of the predictors is assumed to be known and less or equal than that of the filter. This results in a new class of algorithms trading off performance with computational complexity.

I. Introduction. Fast least squares adaptive transversal schemes for multichannel filtering with varying number of memory taps per channel have been derived in [1] - [3]. The common feature of all these algorithms is that for each time update of the unknown filter's impulse response, time updates for the forward as well as the backward multichannel predictors are required. Furthermore the order of these predictors is implicitely assumed to be the same with the order of the filter to be identified. This is a result of the partition properties of the autocovariance matrix. However this is not necessarily true and in many cases the predictors' order is much less than the filter's order. A typical example is that of an echo canceller for the special single channel case. There, the order of the filter may be of the order of thousand, although the input signal can adequately be predicted with orders of 15-20. The aim of this paper is to overcome this apparent contradiction by incoporating into the problem the a priori information about the predictors' order. This idea was first introduced in [4] where the single channel case was treated.

The philosophy behind the proposed method is to assume that the input signal can be described by a multichannel AR model of order m₁, m₂ (AR (m₁, with $m_1 \le i$, $m_2 \le j$, where i, j of filter taps associated with the input As a consequence the Least Squares numbers channels. estimates of the (m₁, m₂) order multichannel predictors will be updated via an available LS time recursive scheme. Thus it is implicitely assumed that the covariance matrix of order $m_1 + m_2 + 2$ is also updated in the usual Least Squares sense, At this point we shall assume that the elements of higher order covariance matrices, can be generated from the above lower order Least Squares counterparts. This is a natural assumption since the input signal has been assumed to be AR (m₁, This way of computing the higher order covariance lags can be considered nonstationary generalization of a maximum entropy extension of the autocorrelation series for an AR process. Each unknown element of the covariance matrix is computed so that to maximize

the minimum prediction error power of the forward or backward predictor of an appropriate order. Exploitation of the properties of the inverse of the resulting matrix leads to an expression relating in a step up step down manner ([1] - [3]) the gain vector of order i,j with the LS predictors of order \mathbf{m}_1 , \mathbf{m}_2 . The proof which is provided is not simply a generalization of the one given in [4] but it follows a different path revealing certain physical insights of the method.

II. Problem formulation. We shall concise ourselves in the two channel case. The idea can obviously be extended to more general multinput multioutput cases.

We are given two input signals $x^1(n)$, $x^2(n)$ and a desired response signal z(n). We seek to compute, in anuoptimal way, the coefficients of an FIR filter

$$y(n) = -\sum_{k=0}^{i-1} c_k^1 x^1 (n-k) + \sum_{k=0}^{j-1} c_k^2 x^2 (n-k)$$
 2.1

so that y(n) to follow z(n) as close as possible. A time iterative solution will be given based on the stochastic Newton algorithm [5].

$$\underline{c}_{ij}(n) = \underline{c}_{ij}(n-1) + \underline{w}_{ij}(n) \xi_{ij}(n)$$
 (2.2a)

where

$$\underline{c}_{ij} \; (n) \!\!=\!\! \left[c_0^1 \; (n), \ldots, c_{i\!-1}^1 \; (n), \; c_0^2 \; (n), \ldots, c_{j\!-1}^2 \; (n) \right]^t \qquad \qquad 2.2\beta$$

$$\boldsymbol{\varepsilon}_{ij}$$
 (n)=[z (n) + \underline{c}_{ij}^{t} (n) \underline{x}_{ij} (n)] 2.2 γ

$$\underline{x}_{ij}$$
 (n)= $[x^1(n) ... x^1(n - i+1) x^2 (n)... x^2(n-j+1)]^{t}$ 2.28

 $\underline{w}_{ij}(n)$ is the gain vector defined as $\underline{w}_{ij}(n)=$ - $R_{ij}(n-1)$ $\underline{x}_{ij}(n)$ with

$$\begin{array}{lll} R_{ij} \; (n) = E \; \left\{ \underline{x_{ij}} \; (n) \; \underline{x}_{ij}^t \; (n) \right\} & 2.2\epsilon \\ We \; shall \; assume \; that \; the \; input \; signals \; are \end{array}$$

generated by a multichannel AR process of order m_1 , m_2 and that estimates of the forward $\underline{A}_{m_1\ m_2}$ (n) and backward $\underline{B}_{m_1\ m_2}$ (n) predictors

$$\begin{bmatrix} \widehat{\mathbf{x}}^{1} & (\mathbf{n}) \\ \widehat{\mathbf{x}}^{2} & (\mathbf{n}) \end{bmatrix} = \underline{\mathbf{A}}_{m_{1} m_{2}}^{t} (\mathbf{n}) \quad \underline{\mathbf{x}}_{m_{1} m_{2}} (\mathbf{n})$$
 2.3\(\alpha\)

$$\begin{bmatrix} \hat{x}^1 & (n - m_1) \\ \hat{x}^2 & (n - m_2) \end{bmatrix} = \underline{B}_{m_1 m_2}^t \underline{x}_{m_1 m_2} \underline{x}_{m_1 m_2} (n)$$
 2.3\begin{align*} 2.3\beta \end{align*}

are available for each time instant: The hat in 2.3 denotes estimates of the corresponding quantities.

III. Extension of $R_{m_1 m_2}(n)$ to $R_{ij}(n)$. This will be caried out recursively. Assume that $R_{pr}(k)$, k=n, n-1,...0 are known. The goal is to compute $R_{p+1 r+1}(k)$, k=n, n-1,...0.

We know that $R_{p,r}(k) = E\left\{\underline{x}_{p,r}(k) \ \underline{x}_{p,r}^{t}(k)\right\}$, $R_{p+l,r+1}(k) =$

$$R_{pr} (k-1) \underline{A}_{pr} (k) = -E \left\{ \underline{x}_{pr} (k-1) \begin{bmatrix} x^{1} (k) \\ x^{2} (k) \end{bmatrix}^{t} \right\}$$
3.1

and the corresponding minimum forward prediction error power trace

trace
$$\left(E \left\{ e_{\mathbf{pr}}^{f}(\mathbf{k}) e_{\mathbf{pr}}^{ft}(\mathbf{k}) \right\} \right) =$$

trace
$$\mathbb{E}\left\{\begin{bmatrix} x^1 & (k) \\ x^2 & (k) \end{bmatrix} [x^1 & (k), x^2 & (k)]\right\} +$$

$$E\left\{\begin{bmatrix} x^1 & (k) \\ x^2 & (k) \end{bmatrix} \underline{x}_{pr}^t & (k-1) \right\} \underline{A}_{pr} (k)$$
 3.2

The unknown plements of $R_{p+1\ r+1}$ (k) will be chosen so that the trace in 3.2 becomes maximum. Consider the following partition

$$\underline{x}_{pr}(k-1) = S_{pr} \begin{bmatrix} \underline{x}_{p-1 r-1}(k-1) \\ x^{1}(k-p) \\ x^{2}(k-r) \end{bmatrix}$$
3.3

where S_{pr} is an appropriately defined permutation matrix. From 3.1 and 3.2 we can get

$$\underline{A}_{pr}(k) = S_{pr} \begin{bmatrix} \underline{A}_{p-1 \ r-1} \ (k) \\ \underline{O}_{2} \end{bmatrix} + S_{pr} \begin{bmatrix} \underline{B}_{p-1 \ r-1} \ (k-1) \\ \underline{I}_{2} \end{bmatrix} k_{pr}^{\frac{2}{3}}(k) \qquad 3.4 \text{ or}$$

where \underline{B}_{pr} (k) is the backward multichannel predictor.

$$R_{p,r}(k) \underline{B}_{p,r}(k) = -E \left\{ \underline{X}_{p,r}(k) \begin{bmatrix} x^{1}(k-p) \\ x^{2}(k-r) \end{bmatrix}^{t} \right\}$$
and
$$K_{pr}^{f}(k) = -\alpha_{pr}^{-b}(k) \left(\underline{B}_{p-1,r-1}^{t}(k-1) \underline{C} + \cdots \right)$$

$$E\left\langle \begin{bmatrix} x^{1} & (k-p) \\ x^{2} & (k-r) \end{bmatrix} \begin{bmatrix} x^{1} & (k), & x^{2} & (k) \end{bmatrix} \right\rangle$$
 3.4 γ

where $\underline{C} = E \left\{ \underline{x}_{p-1 \ r-1} \ (k-1) \left[x^1 \ (k), x^2 \ (k) \right] \right\}$

$$\alpha_{p\ r}^{b}\ (k) = r_{0}^{b}\ (k) + \underline{B}_{p\ r}^{t}\ (k)\ E\ \left\{\underline{x}_{p\ r}\ (k)\ \left[x^{1}\ (k-p)\ x^{2}\ (k-r)\right]\right\} \ 3.48$$

$$r_0^b(k) = E\left\{\begin{bmatrix} x^1(k-p) \\ x^2(k-r) \end{bmatrix} \begin{bmatrix} x^1(k-p) & x^2(k-r) \end{bmatrix}\right\}$$
 3.48

Substituting 3.4 into 3.2 we obtain

$$trace \ \left(E \ \left\{ e_{p \ r}^f \ (k) \ e_{p \ r}^{ft} \ (k) \right\} \right) = trace \ \left(E \ \left\{ e_{p - 1 \ r - 1}^f e_{p - 1 \ r - 1}^{ft} \right\} \right)$$

- trace
$$\left(k_{p\,r}^{ft}(k) \alpha_{p\,r}^{f}(k) k_{p\,r}^{f}(k)\right)$$
 3.5

However $\alpha^f_{pr}(k)$ is a positive definite matrix thus the left hand side of 2.7 achieves its maximum value for $K^f_{pr}(k) = 0$. Thus the unknown elements are given by

$$E\left\{\begin{bmatrix} x^{1} & (k-p) \\ x^{2} & (k-r) \end{bmatrix} \left[x^{1} & (k), x^{2} & (k) \right] \right\} = \frac{B_{D-1, F-1}^{t} (k-1) E\left\{ \underbrace{x_{D-1, F-1} (k-1) \left[x^{1} & (k), x^{2} & (k) \right] \right\}}_{3.6}$$

and this is equivalent to

$$\underline{A}_{p r}(k) = S_{p r} \begin{bmatrix} \underline{A}_{p-1 r-1}(k) \\ O_2 \end{bmatrix} k = n, n-1,...$$
 3.7

If instead of the forward predictor we consider the backward one, similar arguments will lead to

$$\underline{B}_{p r}(k) = T_{p r} \begin{bmatrix} O_2 \\ \underline{B}_{p-1 r-1}(k-1) \end{bmatrix} k = n, n-1,...$$
 3.8

Furthermore the resulting unknown elements are identical with those resulting from the forward predictor. Following the above procedure we can obtain $R_{ii}(n)$

starting from $R_{m_1 m_2}$ (n). It is obvious from the method above that i-m₁ = j-m₂. This is a restriction and we cannot start from an arbitrary couple m_1 , m_2 to reach i,j. The permutation matrix T_{pr} is defined from the following partitioning.

$$\underline{x}_{p r}(k) = T_{p r} \begin{bmatrix} x^{1}(k) \\ x^{2}(k) \\ \underline{x}_{p-1 r-1}(k) \end{bmatrix}$$
 3.9

The resulting covariance matrices can be shown to be positive definite.

IV. The algorithm. In this section the properties of the above extended matrix will be used to derive the novel algorithm. Define

$$\underline{\underline{A}}_{p r}(k) \equiv \begin{bmatrix} \underline{\alpha}_{p}^{11}(k) & \underline{\alpha}_{p}^{21}(k) \\ \underline{\alpha}_{r}^{12}(k) & \underline{\alpha}_{r}^{22}(k) \end{bmatrix}$$

$$4.1\alpha$$

$$\underline{B}_{p r}(k) \equiv \begin{bmatrix} \underline{b}_{p}^{11}(k) & \underline{b}_{p}^{21}(k) \\ \underline{b}_{r}^{12}(k) & \underline{b}_{r}^{22}(k) \end{bmatrix}$$

$$4.1\beta$$

From the definitions in the previous section and the partition properties 3.3 and 3.9 it is shown that

$$\begin{split} R_{i+1\ j+1}^{-1} &= T_{i+1\ j+1} \begin{bmatrix} o & \underline{O}^t \\ \underline{O} & R_{ij}^{-1} \ (n-1) \end{bmatrix} T_{i+1\ j+1}^t + \\ T_{i+1\ j+1} \begin{bmatrix} I_2 \\ A_{ij} \ (n) \end{bmatrix} \alpha_{ij}^{-f} \ (n) \begin{bmatrix} I_2 & \underline{A}_{ij}^t \end{bmatrix} T_{i+1\ j+1}^t & 4.2 \end{split}$$

From 4.2, 3.7, 3.48 and 4.1 α we can show that.

$$R_{i+1}^{-1}_{i+1}(n) =$$

$$T_{i+1 \ j+1} \begin{bmatrix} 0 & \underline{O}^{t} \\ \underline{O} & R_{ij}^{-1} \ (n-1) \end{bmatrix} T_{i+1 \ j+1}^{t} + \underline{\widetilde{A}} \ (n) \ \alpha_{m_{i} \ m_{2}}^{-1} \ (n) \ \underline{\widetilde{A}}^{t} \ (n) \quad 4.30$$

$$\widetilde{A}^{t} \ (n) = \begin{bmatrix} 1 & 0 & \frac{11t}{\alpha_{m_{1}}} & (n) & \underline{O}_{l-m_{1}}^{t} & \frac{\alpha_{m_{2}}}{\alpha_{m_{2}}} & (n) & \underline{O}_{l-m_{2}}^{t} \\ 0 & 1 & \underline{\alpha}_{m_{1}}^{2it} & (n) & \underline{O}_{l-m_{1}}^{t} & \underline{\alpha}_{m_{2}}^{22t} & (n) & \underline{O}_{l-m_{2}}^{t} \end{bmatrix} T_{i+1 \ j+l}^{t} \quad 4.3\beta$$

Also using the backward predictor we obtain

$$\begin{split} R_{1+1\ j+1}^{-1}\ (n) &= S_{1+1\ j+1} \left[\begin{matrix} R_{1j}^{-1}\ (n) & \underline{O} \\ \underline{O}^t & 0 \end{matrix} \right] S_{1+1\ j+1}^t \ + \\ &+ \underbrace{\widetilde{B}}_{1}\ (n-i\ + m_1\)\ \alpha_{m_1\ m_2}^{-b}\ (n-i\ + m_1\)\ \underline{\widetilde{B}}^t\ (n-i\ + m_1\) \ \ 4.4c \end{split}$$

where Bt (k)≡

$$\begin{bmatrix} \underline{O}_{i \cdot m_{i}}^{t} & \underline{b}_{m_{1}}^{11t} (k) & \underline{O}_{j \cdot m_{2}}^{t} & \underline{b}_{m_{2}}^{12t} (k) & 1 & 0 \\ \underline{O}_{i \cdot m_{1}}^{t} & \underline{b}_{m_{1}}^{21t} (k) & \underline{O}_{j \cdot m_{2}}^{t} & \underline{b}_{m_{2}}^{22t} (k) & 0 & 1 \end{bmatrix} S_{i+1 \ j+1}^{t} \quad 4.44$$

 $\alpha_{m_1 m_2}^f(k)$ above is defined as

$$\alpha_{m_{1} m_{2}}^{f}(k) = E \left\{ \begin{bmatrix} x^{1}(k) \\ x^{2}(k) \end{bmatrix} \begin{bmatrix} x^{1}(k) & x^{2}(k) \end{bmatrix} \right\} + \frac{A_{m_{1} m_{2}}^{t}(k)}{E \left\{ x_{m_{1} m_{2}}(k-1) \begin{bmatrix} x^{1}(k), x^{2}(k) \end{bmatrix} \right\} + 4.1\gamma$$

Combining 4.3 and 4.4 with the definition of gain, the following step up - step down recursions are obtained

$$\underline{\mathbf{w}}_{i+1 j+1} (\mathbf{n}) =$$

$$T_{i+1 \; j+1} \; \begin{bmatrix} 0 \\ 0 \\ \underline{w}_{ij} \; (n\!-\!1) \end{bmatrix} + \frac{1}{\lambda} \; \underline{\widetilde{A}} \; \; (n\!-\!1) \; \; \alpha_{m_1 \; m_2}^{-f} \; \; (n\!-\!1) \; \; e_{m_1 \; m_2}^{f} \; \; (n) \qquad 4.5\alpha$$

$$\underline{\mathbf{w}}_{i+1 \ j+1} \ (\mathbf{n}) \ = \ \mathbf{S}_{i+1 \ j+1} \begin{bmatrix} \underline{\mathbf{w}}_{ij} \ (\mathbf{n}) \\ \mathbf{0} \\ \mathbf{0} \end{bmatrix}$$

$$+ \ \frac{1}{\lambda} \ \ \frac{\widetilde{B}}{B} \ (n-i+m_1-1) \ \ \widetilde{\alpha}_{m_1 \ m_2}^b \ (n-i+m_1-1) \ \ e_{m_1 \ m_2}^b \ (n-i+m_1) \ \ \ 4.5\beta$$

where $e_{m_1\ m_2}^f$, $e_{m_1\ m_2}^b$ the a priori forward and backward errors of order $m_1\ m_2$.

Finally we shall obtain the relation between $\mathbf{\epsilon}_{ij}(n)$ defined in 2.2 γ and the corresponding prior error $\mathbf{e}_{ij}(n)$. This can easily be shown to be

$$\epsilon_{ij}(n) = \frac{e_{ij}(n)}{1 - \alpha_{ij}(n)}$$
4.6 α

where

$$\alpha_{ij}(n) = \sum_{ij}^{t}(n) w_{ij}(n)$$
 4.6 β

From 4.4 and 4.5 we obtain the following.

$$\alpha_{ij} \ (n) \ = \ \alpha_{ij} \ (n\text{-}1\) \ + \ \frac{1}{\lambda} \ e^{ft}_{m_1 \ m_2} \ (n) \ \alpha^{\text{-}f}_{m_1 \ m_2} \ (n\text{-}1) \ e^{f}_{m_1 \ m_2} \ (n)$$

$$-\frac{1}{\lambda} e^{tt}_{m_1 m_2} (n+m_1-i) \alpha^{-b}_{m_1 m_2} (n+m_1-i-1) e^{b}_{m_1 m_2} (n+m_1-i) \qquad 4.7$$

Combination of 2.2 α , 4.5 α , β , 4.5 γ , 4.6 and 4.7 gives the new algorithm. The complexiby is 2 (i + j) + 5(m_1+m_2). If the input model is a true AR (m₁ m₂) then by increasing (m₁ m₂) to (i j) there is no gain in performance. It however this is not true, then varying (m₁ m₂) from a minimum value to (i j) a whole range of algorithms results trading off performance with computational complexity.

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